



Issues in Mobile Multi-User MIMO Networks

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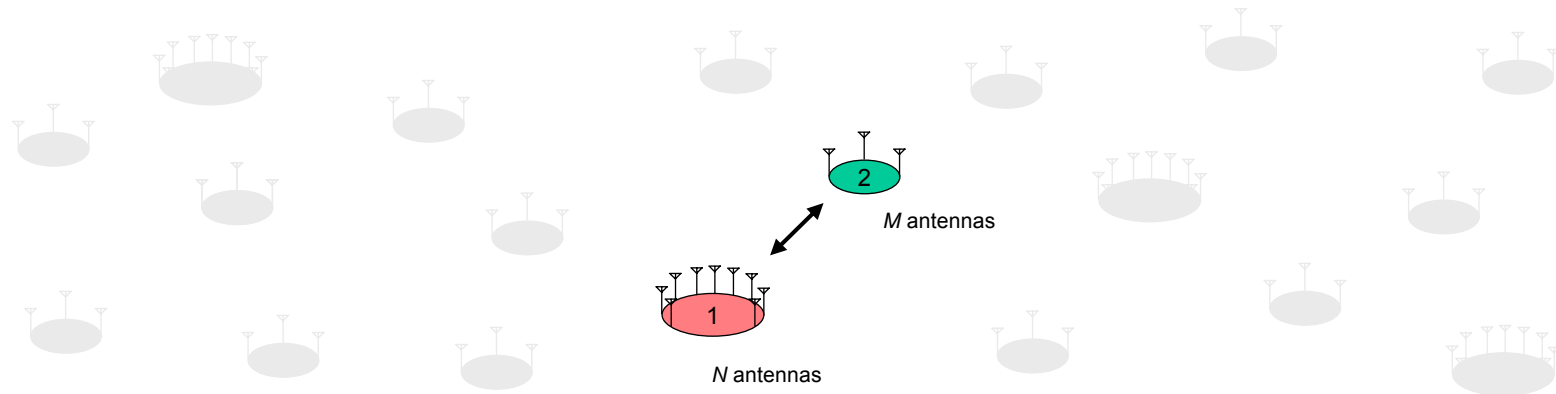




Presentation Outline

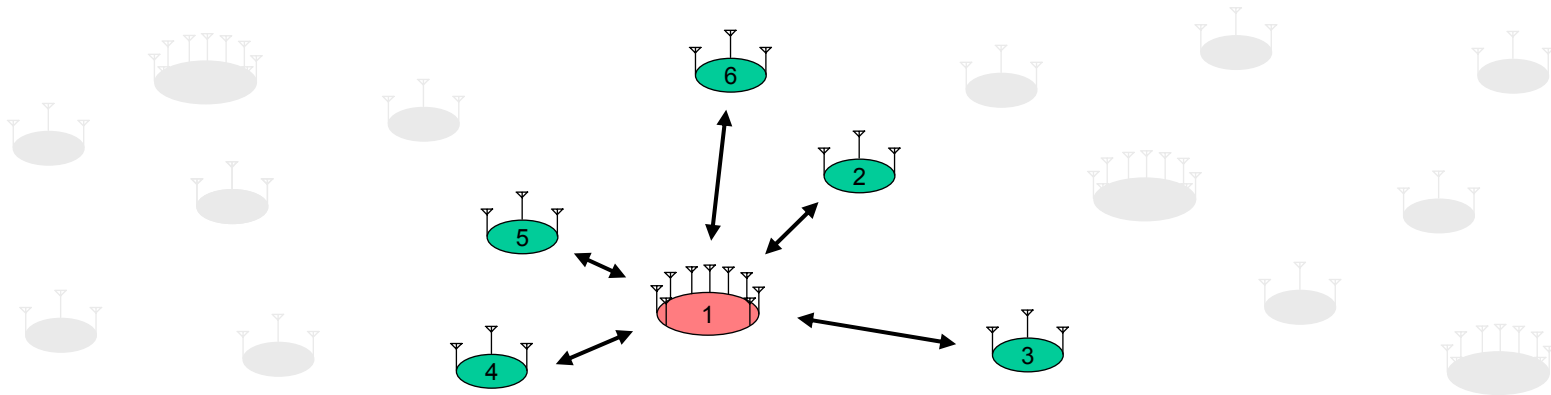
- Channel State Information
 - Single-User vs. Multi-User MIMO Scenarios
 - Multicast, Broadcast Examples
- Coordinated Tx-Rx Beamforming
 - Issues
 - Alternatives
- Interference-Dependent Coding
 - Channel Inversion
 - Modulo Pre-coding
- MIMO Channel Prediction
- Other Interesting Directions

Channel State Information: Single-User, Point-to-Point MIMO



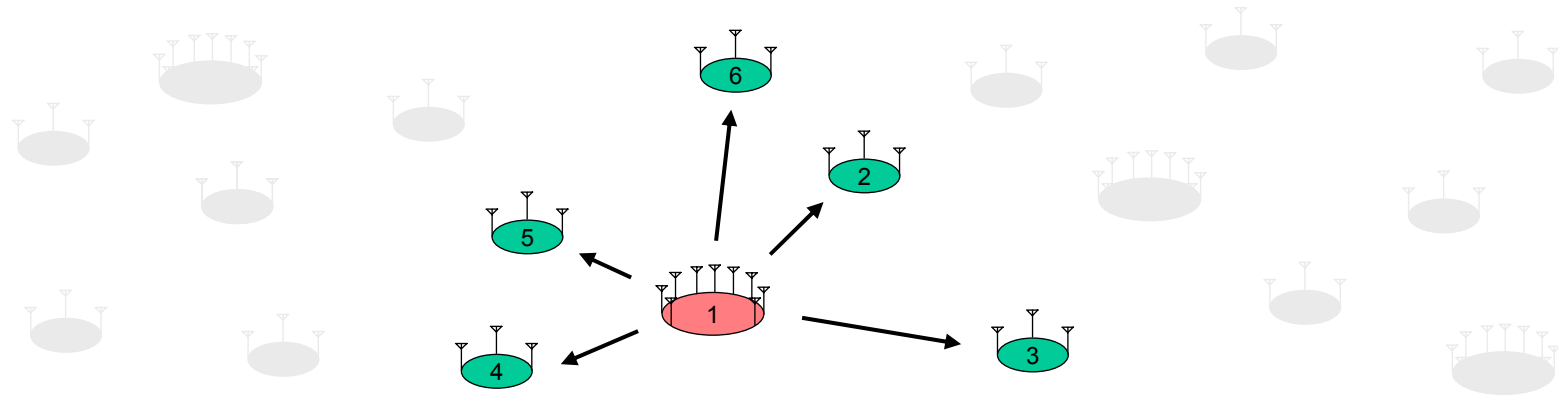
- Under ideal conditions (e.g., independent Rayleigh fading)
 - capacity grows linearly with $\min(N, M)$
 - capacity growth is independent of CSI at the transmitter
- If channel is rank deficient or $N > M$, transmit CSI much more important
- Single-user point-to-point model appropriate if all network neighbors are time- or frequency-multiplexed.

Channel State Information: Multi-User Network MIMO



- Spatial multiplexing possible if node A is multicasting or “multi-routing”
- Transmit CSI is now critical to balance interference & throughput
 - receivers may not be able to perform MUD
 - little capacity growth with N , M or SNR
- Intelligent allocation of spatial resources also requires transmit CSI

MIMO Multicast Example



Channels: $\mathbf{H}_2, \mathbf{H}_3, \mathbf{H}_4, \mathbf{H}_5, \mathbf{H}_6$

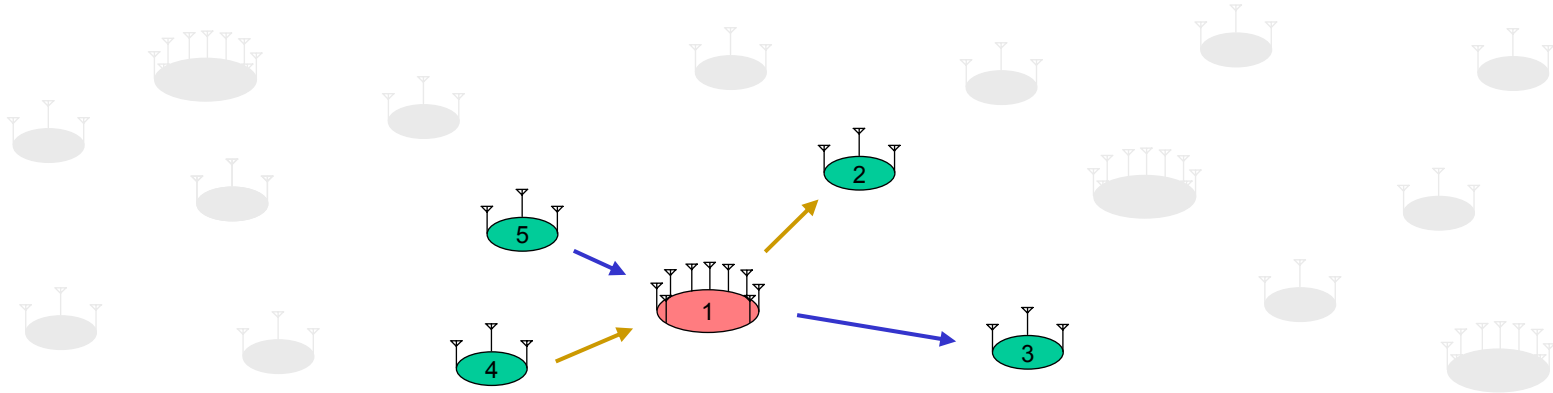
Transmit beamformer: \mathbf{b}

Receive data model: $x_2 = \mathbf{w}_2^* \mathbf{H}_2 \mathbf{b} d + n_2 \quad x_3 = \mathbf{w}_3^* \mathbf{H}_3 \mathbf{b} d + n_3 \quad \dots$

Receive beamformers: $\mathbf{w}_2 = \mathbf{H}_2 \mathbf{b} \quad \mathbf{w}_3 = \mathbf{H}_3 \mathbf{b} \quad \dots$

Problem: $\min_{\mathbf{b}} \mathbf{b}^* \mathbf{b} \quad \text{s.t.} \quad \mathbf{b}^* \mathbf{H}_i^* \mathbf{H}_i \mathbf{b} \geq \gamma_i, \quad i = 2, \dots, 6$

Spatially Multiplexed Routing Example (MIMO Broadcast Channel)



Channels: $\mathbf{H}_2, \mathbf{H}_3, \mathbf{H}_4, \mathbf{H}_5$

Transmit beamformer: $\mathbf{b}_2, \mathbf{b}_3$

Receive beamformers: $\mathbf{w}_3 = (\mathbf{H}_2 \mathbf{b}_2 \mathbf{b}_2^* \mathbf{H}_2^* + \sigma^2 \mathbf{I})^{-1} \mathbf{H}_3 \mathbf{b}_3$

Problem:

$$\min_{\mathbf{b}} \mathbf{b}_2^* \mathbf{b}_2 + \mathbf{b}_3^* \mathbf{b}_3 \quad \text{s.t.} \quad \frac{|\mathbf{w}_i^* \mathbf{H}_i \mathbf{b}_i|^2}{|\mathbf{w}_i^* \mathbf{H}_i \mathbf{b}_j|^2 + \sigma^2 \mathbf{w}_i^* \mathbf{w}_i} \geq \gamma_i, \quad \{i, j\} = 2, 3$$



Coordinated Tx-Rx Beamforming

- despite non-convex constraints, multicast beamforming problem is efficiently solved via semi-definite optimization (SDO)
- unless receive nodes have 1 antenna, this is *not* true for broadcast case
- possible iterative solution:
 - (i) assume initial set of receive beamformers \mathbf{w}_i
 - (ii) find optimal transmit beamformer \mathbf{b} using SDO
 - (iii) recalculate receive beamformers
 - (iv) go to step (ii)
- Issues:
 - when can convergence be guaranteed?
 - are there simpler alternatives to SDO for step (ii)?
 - are there alternatives to full transmit CSI?
 - optimal initialization to reduce # of iterations
 - sensitivity to CSI accuracy
- Alternative strategies:
 - downlink training, uplink “best” single-stream beamformer
 - beamformer “codebooks”
 - interference-dependent (“dirty-paper”) coding
 - MIMO channel prediction



Interference-Dependent Coding

- although relatively simple, linear beamforming is not capacity optimal
- “dirty-paper” techniques are known to achieve sum capacity in certain cases
- a broadcast example:

xmit node has M antennas, K rcv nodes w/ 1 antenna each: $\mathbf{x} = \mathbf{H}\mathbf{s}$
 $K \times 1$ $K \times M$

channel inversion: $\mathbf{s} = \gamma \mathbf{H}^* (\mathbf{H}\mathbf{H}^*)^{-1} \mathbf{d}$

performs poorly even in ideal i.i.d. Rayleigh fading case
ill-conditioned channel leads to signal attenuation, not noise amplification

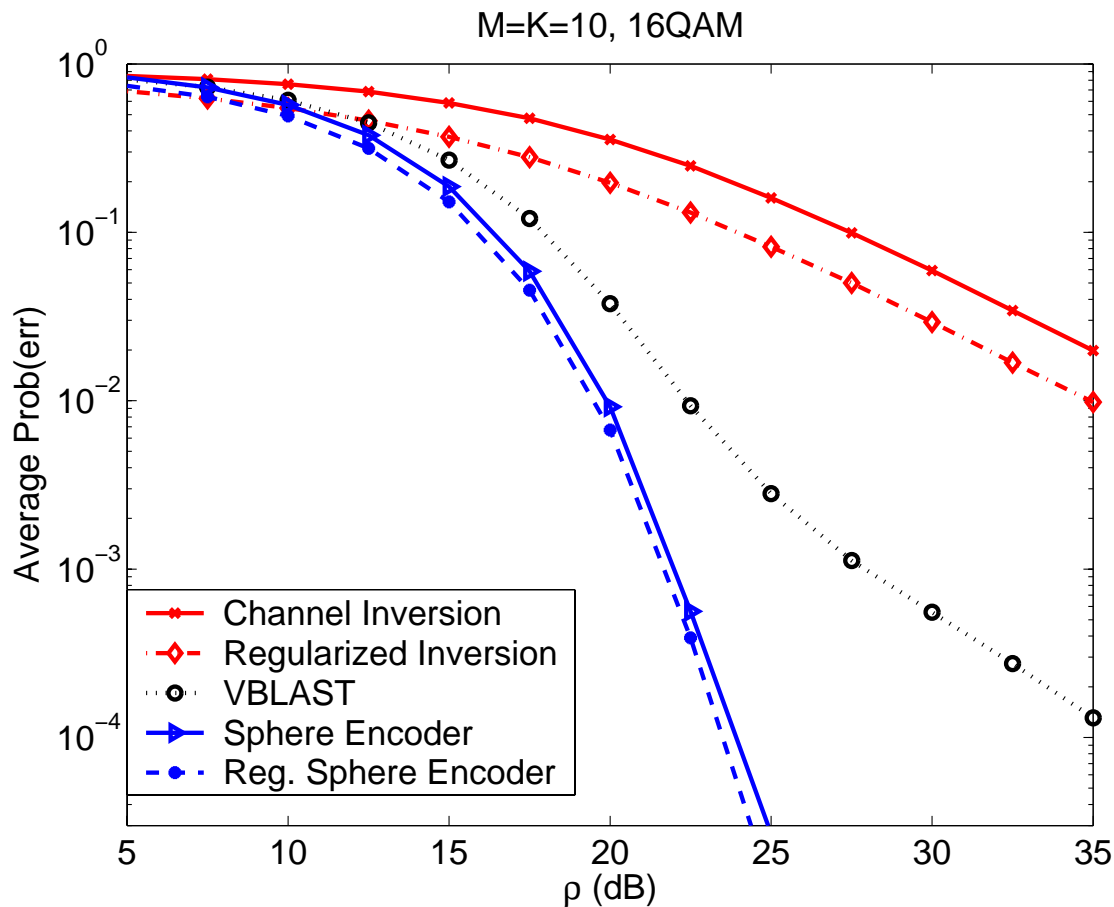
regularized channel inversion: $\mathbf{s} = \gamma \mathbf{H}^* (\mathbf{H}\mathbf{H}^* + \alpha \mathbf{I})^{-1} \mathbf{d}$

offers some improvement, but must choose α carefully

modulo precoding: $\mathbf{s} = \gamma \mathbf{H}^* (\mathbf{H}\mathbf{H}^*)^{-1} (\mathbf{d} + \tilde{\mathbf{d}})$

code perturbation is channel (interference) dependent
perturbation is restricted to integer lattice for modulo decoding
can be thought of as “sphere encoding”

Modulo Precoding Example

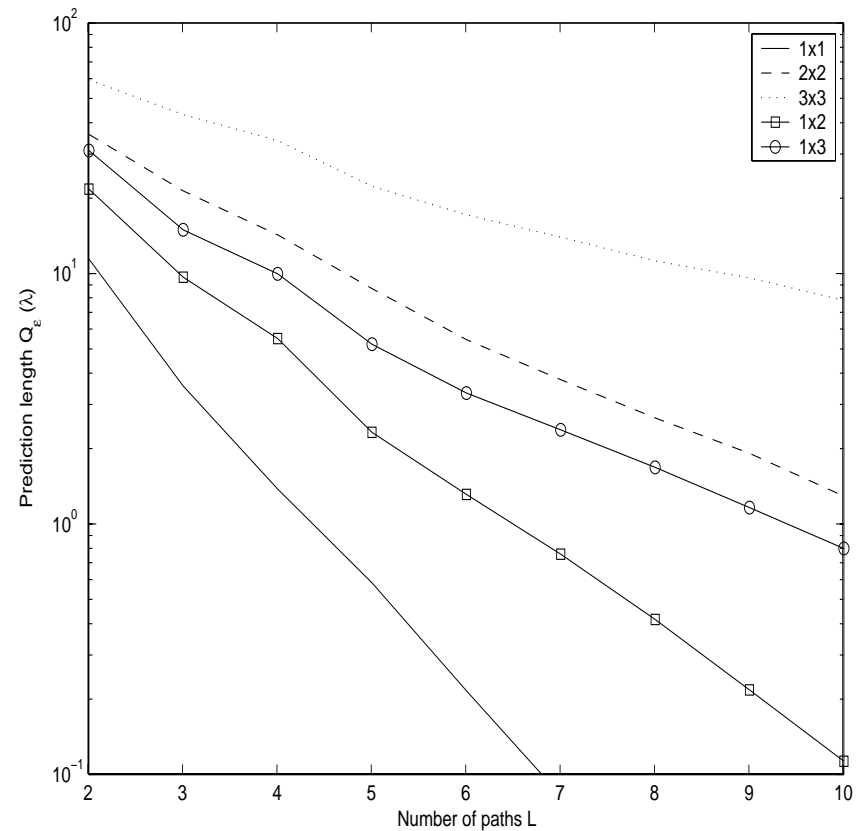
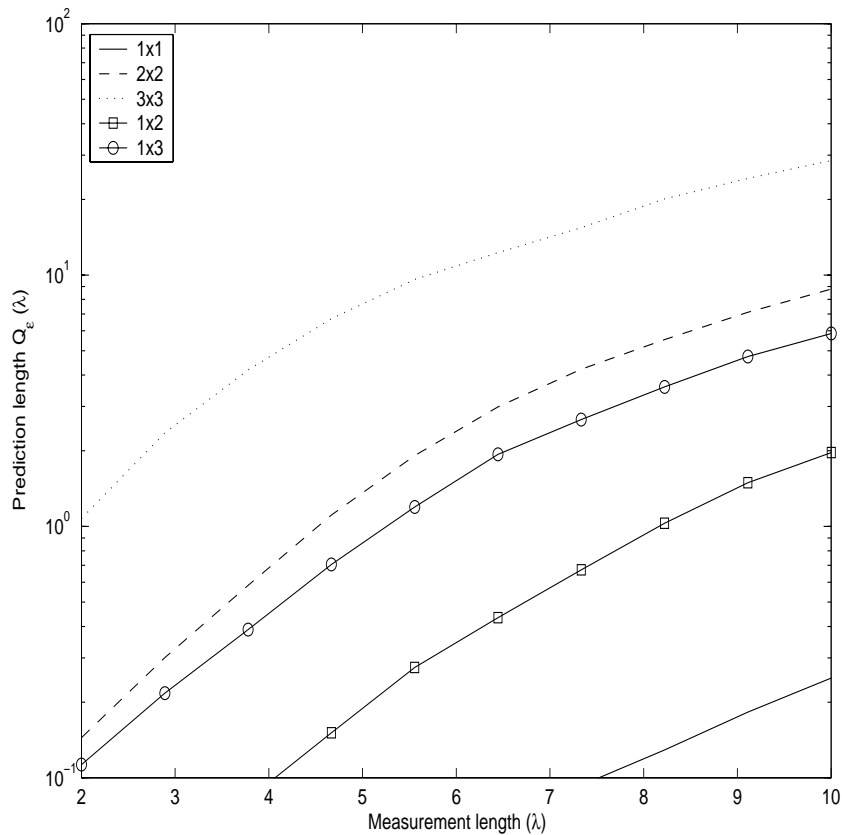


Issues:

- what if rcv nodes have more than 1 antenna?
- no analytical performance analysis now available
- better performance can be achieved with higher dimensional lattices
- computationally complex

MIMO Channel Prediction

CRB analysis indicates that channel prediction horizons are much longer for MIMO than SISO systems (joint work w/ T. Svantesson):



MIMO Channel Prediction Issues

- algorithms to realize the bounds (or get close)
- Kalman filters, auto-regressive models are obvious possibilities
- element-wise AR models do not capture MIMO structure

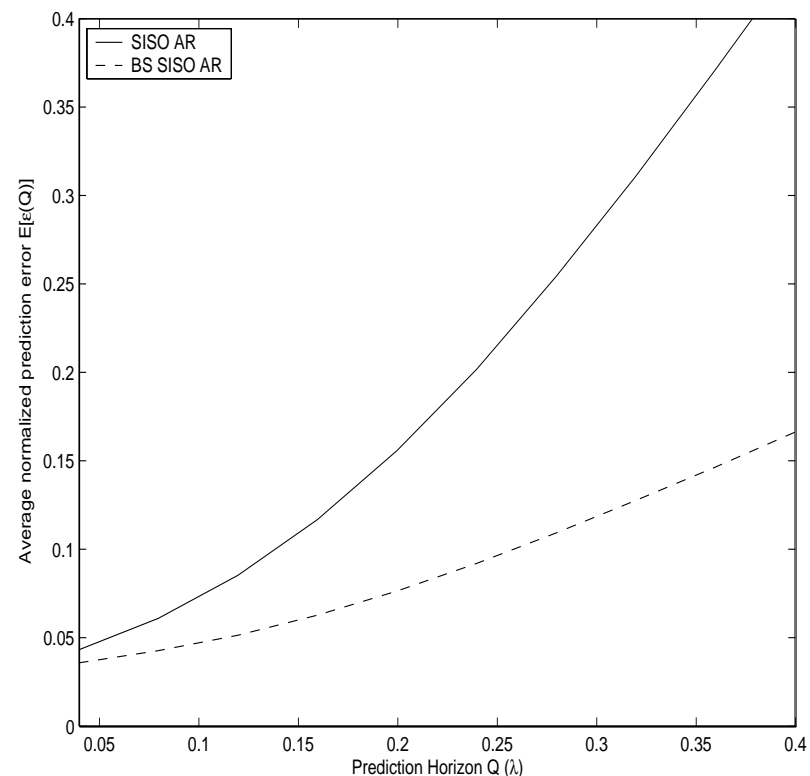
$$h_{ij}(t) = \alpha_{ij}h_{ij}(t - 1) + \varepsilon(t)$$

- alternative: beamspace AR

$$\bar{h}_{ij}(t) = \alpha_{ij}\bar{h}_{ij}(t - 1) + \varepsilon(t)$$

$$\bar{\mathbf{H}}(t) = \mathbf{W}^* \mathbf{H}(t) \mathbf{W}$$

- example shows results for 10x10 ULA and uniform DOA sampling
- performance still significantly short of the CRB
- improvements possible using channel-dependent beamspace transformation



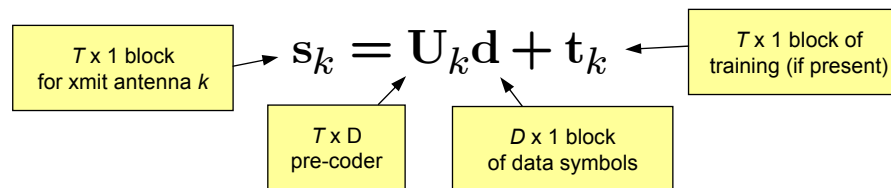
Other Interesting Directions

- blind & semi-blind channel estimation

- reduces training overhead
- exploit structure due to space-time coding

- beamforming vs. space-time coding trade-off

- both are special cases of affine pre-coding framework:



- all rational rates possible, determined by pre-coder dimensions
- allows for superimposed training
- prior work exploits framework for (semi-)blind equalization

- resource allocation

- multi-user diversity, routing decisions require single FOM for each channel
- due to spatial dimension, there are multiple FOMs per user
- FOMs depend on which users are spatially multiplexed
- need “spatial compatability” metrics; e.g.,

$$\xi_{ij} = \frac{\|\mathbf{H}_i^* \mathbf{H}_j\|_F^2}{N_i N_j}$$